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PSYCHOACOUSTICS

In this article we apply signal detection theory to qualitatively unify the intensity just-noticeable difference (JND) and masking data of Wegel and Lane (1), Fletcher and Munson (2,3,25,31), Miller (4), and Egan and Hake (5). We do this by treating the loudness as the first moment, and the intensity JND as the second moment, of the random variable we call the *single-trial loudness*. From these definitions we define a loudness signal-to-noise ratio $(SNR_{\mathscr{D}})$, which is the unifying factor. This theory relies heavily on Fletcher and Munson's 1933 theory of loudness (2). The purpose of this article is to create a model of masking that may be used for speech and music coding.

Introduction

The Problem of Perceptual Coding. When quantizing signals one necessarily, by design, introduces noise into the representation. The art of *perceptual coding* is to control this noise so that it has the smallest perceptual effect. Given a complete description of the signal-dependent internal noise of the auditory system, it is assumed that it is possible to quantize the signals with a similar noise. Given such a coder, the quantizing error would be perceptually *masked* by the internal noise of the auditory system.

When the quantizing error is large enough that the error is above the perceptual threshold, we would like the system to degrade gracefully. How can we meet these difficult goals? The short answer is: only with a model of masking.

What is masking and where does it come from? How do we measure it experimentally? How may we predict it for an arbitrary signal? To understand the answers to these questions, we need models of loudness, masking, the intensity just-noticeable difference (JND_I) , critical bands, and the listening condition. This article is about modeling masking and is an attempt to describe masking to the designer of a coder. It is not about how to engineer a speech or music coder.

When dealing with human perceptions we must carefully distinguish the *external* physical variables, which we call Φ variables, from the *internal* psychophysical variables, which we call Ψ variables (note that Φ and Ψ sound like the initial syllable of the words *physical* and *psychological*, respectively). The model we seek is a transformation from the Φ domain to the Ψ domain. Examples of Φ variables are pressure, frequency, intensity, and neural rate. Neural rate is an important example because it is an internal variable, yet it is physical; thus the terms *internal* and Ψ are not synonymous. Example of Ψ variables are loudness, pitch, heaviness, brightness, tightness, and timbre.

To represent an acoustic signal, many Φ variables (e.g., intensity, frequency, duration modulation) must be specified. In a similar manner, to understand and model masking, many internal Ψ variables must be determined. There is a natural but unfortunate tendency (everybody does it, all the time), to confuse the external physical (Φ) variables with the internal psychophysical (Ψ) variables (for example: heaviness and weight, pitch and frequency). Such confusions, in a scientific context, lead to serious misinterpretations of experimental results.

While the physical variables may be either deterministic or stochastic, it is essential to treat Ψ variables as stochastic. In fact, it is the stochastic character of the Ψ variables that is responsible for masking. Furthermore, to understand masking completely, it is important to describe it in both domains. Given these models, the Ψ -domain noise may be reflected back to the external Φ domain, into the acoustic domain of the speech or music signal. For example, in a coder design application, we must first completely characterize the auditory Ψ -domain signal-to-noise ratio (e.g., the ratio of loudness to loudness noise), and then describe it in terms of the Φ -domain signal-to-noise ratio (e.g., the ratio of the ear canal pressure or intensity to JND_I).

It is highly recommended that the serious reader carefully study Chaps. 10, 11, and 13 and Appendix D of Yost (6). Other important general sources are Gelfand (7) and the excellent book of Littler (8), which, along with Fletcher (9), was a primary source of information for the author.

Some History. By 1918 AT&T had decided that if they were going to transmit speech across the country by wire, it was essential that they understand the physical processes behind auditory communication. This was clearly articulated in an internal report written in 1920 by J. Q. Stewart of the AT&T Development and Research (D&R) Department, which oversaw the funding for the Western Electric Engineering group:

The desirability of making direct studies of the physics and psychology of hearing in the quality investigation is becoming increasingly more evident. Research on the physical nature of speech alone will not be sufficient to establish the physical basis for the prediction of articulation; but it must be supplemented by studies of hearing. Indeed, the latter seem to be even more important, because the progress which already has been made toward the formulation of a general philosophy of articulation has not been dependent to any degree on knowledge of the physical characteristics of speech, but has been dependent on hypothesis relating to hearing.

This set the tone for the next 27 years of work. By 1922 a research project was in full gear due to the work of Harvey Fletcher, who had joined AT&T's Western Electric engineering department in 1916. Fletcher's Ph.D. concerned what is now known as the "Millikan oil drop experiment," and AT&T hired him, based on his basic understanding of the physics of the electron, to help build a better telephone system. Fletcher quickly became the leader of this large and important research effort, and by 1924 AT&T was fully committed to funding basic research on speech perception. This was a true team effort, led by Fletcher.

Fletcher and Wegel (10) accurately measured (for the first time) the threshold of hearing. The year after their study, Knudsen, a student of Fletcher's, was the first to measure the pure-tone intensity JND (11). Then Fletcher (12), with the help of Wegel and Lane (1), provided critical and detailed tone-on-tone masking data. Fletcher and Steinberg then showed the relation between Wegel and Lane's masking data and partial loudness (13). In 1924 Wegel and Lane outlined the physical theory of the cochlear traveling wave (1). Kingsbury measured isoloudness contours that defined the *loudness-level*, or *phon*, scale (14;15, p. 227). Finally, Riesz (16) conducted an extensive study of tonal masking for probes near the masker frequency (6,8).

This series of theoretical and experimental studies showed that cochlear filtering results from a traveling wave on the basilar membrane, and the power-law nature of loudness growth in the cochlea was established. Fletcher summarized this work in the first book to be written on speech and hearing (15), bringing him worldwide acclaim.

In 1933, based on detailed loudness and masking data of Munson and on a theory worked out by Fletcher, Fletcher and Munson published the first model of loudness (2). They described how the auditory signal is broken down into many frequency bands and compressed by the cochlea prior to being neurally coded. They described partial loudness as the neural rate, provided the functional relationship between masking and partial loudness, and showed how partial loudness is summed to given the total loudness (17). Unfortunately, while this paper was highly regarded, the information was not widely digested and accepted by the research community. For example, the idea that tonal loudness is additive is still controversial (18).

By 1935, it was proposed that the randomness in the neural representation is what makes the Ψ variables random variables (19,20). Based on the nature of hearing-loss and loudness recruitment, it became clear by 1937 that the basilar membrane signal is compressed, that this compression is lost when the cochlea is damaged (21), that this cochlear damage is due to the loss of hair cells (22,23), and that the resulting loss of outer hair cells causes loudness recruitment (17). At high intensities, all the neurons are stimulated and the loudness is approximately the same as in the normal ear; however, because of the loss of the nonlinear compressive action in the recruiting ear of the outer hair cells, the dynamic range of the cochlea is reduced.

By 1947 the similarity between the JND and masking had been clearly articulated and quantified (4). However, the significance of the neural noise proposal of Montgomery was not appreciated for at least 30 years, when Siebert (24) applied signal detection theory to various JND data. As will be described below, while neural noise appears to determine the limits of the pure-tone JND task under many conditions, it may not the limiting factor in all listening tasks. If another uncertainty (e.g., external noise) dominates the neural uncertainty, then that factor will limit the JND.

Summary. Fletcher's 1933 loudness model is basic to our present understanding of loudness coding (2). With this model Fletcher had clearly described tonal loudness in terms of an additive neural rate code. While we still need to fill in many details, in my opinion this basic idea (however controversial), is correct. By 1950 it had been extensively tested (3,9,25,26).

From the loudness model, several predictions are important. These include the auditory threshold, which is modeled as zero loudness (2), the JND, which provides an inferred measure of the loudness uncertainty (i.e., "internal noise") (4,19,27,28), the masked threshold, which is a generalized JND measure of signal uncertainty (2,29), signal duration effects (30), and the critical band (25,31). Many other phenomena (e.g., the frequency JND, beats) appear to be accounted for by these models (9), but the details need to be reevaluated in terms of the theory of signal detection (TSD).

What is seriously lacking in our present-day understanding of auditory signal processing is a quantitative model of masking. While Fletcher's loudness model shows us how to calculate the loudness given the masked threshold (2,3,25,32) and provides detailed experimental results on the relations between masking and loudness of tones and noise of various bandwidths (9), he did not (nor did Zwicker) provide us with an accurate method for the direct calculation of the masking pattern for arbitrary signals. Reasons for this include the lack of a quantitative understanding (i.e., models) of (a) the highly nonlinear, upward spread of masking (1), (b) the stochastic nature of masking (19), and (c) beats. [Fletcher was also busy working out his empirical theory of speech perception (33–34,35).] Thus masking models are the key to understanding hearing, improving speech coding, and quantifying models of loudness. Masking is one of the most important topics in psychophysics, yet it is arguably the most poorly understood.

During the 1960s the theory of signal detection (TSD) was identified as an important tool for quantifying the detection of probe signals. What is missing is detailed models of the decision variables used by the central nervous system (CNS). For the JND case, where the probe is a copy of the masker, the relevant decision variable is the change in loudness (17,27,36). In the case of masking, where the probe and masker are different, the decision variable is unknown and must be experimentally deduced with a model and experimental data for a multiplicity of maskers and probes. To model the masked threshold (i.e., masking) we need more than a model of loudness: we need a model of the probe signal-detection task. This model must start with the partial loudness of the masker plus probe, and describe the detection probability of the probe signal, as a function of the probe and masker physical parameters.

Overview of this Article. In the next section we will define important concepts such as loudness, the intensity JND (sometimes called "self-masking"), masking, and critical bands. The discussion in that section is limited to describing the definitions. Readers familiar with these definitions may move directly to the analysis and discussion in the subsequent sections on "The Loudness Signal-to-Noise Ratio" and "Narrowband Maskers," and use the "Definitions" section as a reference.

We shall argue that masking is synonymous with the uncertainty of the Ψ representation. In engineering jargon, masking results from the "quantizing noise" of the loudness representation. Because of this relationship, and its simplicity, the JND has special significance to the theory of masking. To analyze the problem in more detail we need to make some distinctions about *n*-interval force choice (*n*-IFC) methods, intensity modulation detection, pure-tone and "frozen-noise" maskers of various bandwidths, and forward masking. Finally we describe a time-domain nonlinear model of the cochlea and auditory system that explains these data and discuss how one might model masking by arbitrary signals, such as music and speech.

When representing signals in a computer there are two basic models: fixed-point and floating-point. In the telephone industry there is a third standard called μ law. In a fixed-point representation the noise is fixed at one-half the least-significant bit (*LSB*). Thus the noise is, to a first approximation, independent of the signal level, and the signal-to-noise ratio (*SNR*) is proportional to the signal. In a floating-point representation, the noise is a fixed percentage of the signal level. For example, with an 8 bit mantissa, the noise floor would be approximately $8 \times 6 = 48$ dB below the signal. Thus the SNR is roughly independent of the signal level. In a μ -law signal the noise depends on the signal level. The ideal μ -law device is similar to a logarithmic compression function and provides a floating-point signal representation, with a 38 dB dynamic range, over the signal range from -40 dB to 0 dB, and a fixed-point representation with a constant noise level over the signal range from -80 dB to -40 dB (36a).

We shall see that in the auditory system, the representation noise is a function of the signal level in a way that is similar to μ -law coding. At low intensities the loudness SNR improves, and at higher intensities it saturates, with a maximum of 30 to 40. Because of the compression of the loudness, specified by the loudness power law, the ratio of the loudness SNR to the signal SNR equals the exponent in the loudness power law. For wideband signals at moderate to high intensities the compression is approximately the square root of the pressure (fourth root of intensity). This means that a loudness SNR of \approx 40, referred to the input signal domain, is \approx 10 (i.e., 10 log₁₀ 10 = 10 dB).

Definitions

The definitions in this section are summarized in Table 1.

Loudness. The *loudness* \mathscr{D} is the name of the Ψ intensity corresponding to an acoustic Φ signal. One of the major conclusions of modern psychophysics is that Ψ variables are random variables (or processes). We define the *single-trial loudness* \tilde{l} as the Ψ random processes that represents the loudness corresponding to each presentation of the signal (the tilde is used to indicate a random variable; all Ψ variables are represented in calligraphic font). A *trial* is defined as a stimulus *presentation* followed by a subject *response*. The presentation can be a single interval, as in loudness scaling methods, or multiple presentations (e.g., an *ABX* is an example of a 3-IFC method; signals A and B are first presented, followed by X, which is either A or B) (6). Since the Ψ variables are random, the subject responses must be random, with a degree of randomness (variance) that depends on the task.

The expected value of $\tilde{l}(I)$,

$$\mathcal{L}(I) \equiv \mathscr{E}[\tilde{\mathcal{L}}(I)]$$

characterizes the loudness $\mathcal{D}(I)$ at intensity *I*, while the variance

$$\sigma^2_{\mathcal{L}}(I) \equiv \mathscr{E}[(\tilde{\mathcal{L}}(I) - \mathcal{L}(I))^2]$$

Quantity	Symbol	Definition
Intensity	I	Eq. (6) or $\mathscr{E}[\tilde{I}]$
Intensity detectoroutput	Ĩ	Eq. (10)
Single-trial loudness	Ĩ	
Loudness	L	8 [Ĩ]
Loudness s.d.	$\sigma_{\mathcal{Z}}$	$\mathscr{E}[(\widetilde{\mathscr{Z}}(I) - \mathscr{L}(I))^2]$
Intensity s.d.	σ_I	$\sigma_{\mathcal{L}} d \mathcal{L} dI $
JNDI	ΔI	Intensity JND
JND 🖉	$\Delta \mathscr{L}$	Loudness JND
SNRI	I/σ_I	Intensity SNR
SNR 🖉	$\mathcal{A}\sigma_{\mathscr{I}}$	Intensity detectorSNR
SNR	$\mathscr{L}\sigma_{\mathscr{L}}$	Loudness SNR
d-prime	d'	Eq. (11)
Weber fraction	J	$\Delta I/I$
Probe gain	α	Probe scale factor
Decibellevel	β	$10 \log_{10}(I/I_{ref})$
Log loudness	-Clog	10 log10 - C
Loudness exponent	γ	$d \mathscr{L}_{ m log}/deta$
	$\widetilde{\Delta} \mathscr{I}(t, \alpha)$	$\widetilde{\mathscr{I}}(t, \alpha) - \widetilde{\mathscr{I}}(t, 0)$
	$\Delta \mathcal{I}(t, \alpha)$	$\mathcal{I}(t, \alpha) - \mathcal{I}(t, 0)$
Threshold probe gain	α^*	$d'(\alpha^*) = 1$

Table 1. Definitions

characterizes the loudness JND $\Delta \mathscr{D} = d' \sigma_{\mathscr{D}}$, where the proportionality factor d' depends on the subject's criterion and on the experimental design. Loudness is an important example of an auditory Ψ variable; without a model of loudness, it is *not* possible to relate (i.e., model) many different experimental results, because the subject's responses are a function of the loudness. A model of the loudness of simple sounds, such as tones and narrow bands of noise, is critical to the theory of masking because the masking is directly related to the loudness *standard deviation* (s.d.) $\sigma_{\mathscr{D}}$.

An Example. Without loudness we cannot make Ψ models of the signals we are processing. An example should help clarify this point. In 1947 Munson studied the question of the effects of signal duration on tonal loudness (30). He matched the loudness of a reference tone at frequency $f_{ref} = 1$ kHz, duration $T_{ref} = 1$ s, and intensity $I^*_{ijk}(I_i, f_j, T_k)$, to target tones of intensities $I_i = 30$ to 90 dB *SPL*, frequencies $f_j = 0.25$ to 10 kHz, and durations $T_k = 5$ to 500 ms. For example, if $\mathscr{D}(I, f, T)$ is the loudness of a tone at intensity I, frequency f, and duration T, then the *loudness level* I^*_{ijk} , in phons, is defined by the relation

$$\mathcal{L}(I_{ijk}^*, f_{ref}, T_{ref}) = \mathcal{L}(I_i, f_j, T_k)$$

Munson was able to model these experimental results accurately by first transforming the matched intensity I^*_{ijk} to a *loudness rate* $\mathscr{D}_t(t)$ per unit time, using the power-law relation of Fletcher and Munson (2), and then integrating the results with a low-pass filter $E_s(t)$, which he called the *sensation integral*. The form of $E_s(t)$ was determined from the data so as to best fit all his experimental results. He then *predicted* the loudness matches to tones having duration of up to 1 min measured previously by von Bekesy. The key to this model is the nonlinear transformation to the loudness variable, and an integration of the resulting loudness rate over time. The additivity of loudness between ears and across frequency had previously been demonstrated in Ref. 2, and is a key property of these loudness models.

Critical Band. The threshold of a pure tone in wideband noise is determined by the loudness standard deviation $\sigma_{\mathscr{D}}$ and the bandwidth of the filters, as measured by the *equivalent rectangular bandwidth* (*ERB*), or critical bandwidth. The *critical band* is the name Fletcher gave to the band-pass filters of the cochlea. When a tone centered on the center frequency of a band-pass filter is presented in a wide band of noise, the filter removes the noise outside of the passband of the filter. Thus the loudness SNR at the output of the filter depends critically on the bandwidth of the filter. This means that the detectibility of the tone in noise is determined by the filter's equivalent power bandwidth (ERB). Since masking is the increase in the threshold of a tone, due to the presence of noise, the critical bandwidth plays an important role when using wideband maskers.

Modeling Loudness. The basic transformation to loudness is a several-step process. First, the input signal is filtered by the cochlea. Second, the dynamic range of the signals is compressed by the outer hair cells on the basilar membrane (17,28,36). Third, the signals are encoded by the inner hair cells and neurons, and a stochastic neural representation results. Masking and partial loudness are formed from this stochastic neural representation, and there is extensive evidence that these two quantities are functionally related (9,37). Finally the neural representation is processed by the nervous system and the total loudness is evaluated. It was proposed by Fletcher and Munson in 1933, for simple signals such as tones and noise, that this final processing is a simple sum of the neural rate (2). While this assumption of partial loudness additivity has been controversial, it has held up amazingly well (18). An analogy that seems appropriate is Newton's apple. Newton could never prove that the apple would always fall, but it always did. Only with the discovery of quantum mechanics were the important limits of Newton's law F = ma uncovered. It is probably true that partial loudness does not always add. But under the conditions of these simple experiments with tones, it always does. Thus it is important to appreciate both the limitations and the power of the additivity of partial loudness (37).

The two models that have described the above steps are those of Fletcher and Zwicker. Both of these models are described in the frequency domain. Both models are deterministic. Loudness however is a random variable. We shall show that after a transformation of tone and noise intensity JND data into the loudness domain, the JND data are greatly simplified.

The emphasis in this article is on predicting masking for arbitrary signals. To do this we need a timedomain loudness model. The problem with all loudness models (including the present one) is the lack of detailed understanding, and therefore specification, of the transformation between the ear canal pressure and the motion of a point on the basilar membrane. This transformation has two components: a linear filtering component and a compressive (nonlinear) component. While approximate solutions to this problem have been proposed, there is presently no accepted model of the compressive transformation. Until such a theory is more fully developed, we must continue to deal directly with experimental data (38). The greatest simplification and understanding of these experimental data is found in a description of the data in terms of *masking patterns* and *partial loudness excitation patterns*. As we have refined our understanding of the nonlinear excitation pattern model, we have been able to account for diverse types of experimental data. However, we cannot consider this problem solved until the physics of the nonlinear transformation, involving the outer hair cells of the cochlea, is fully described.

Loudness Growth. Φ intensity is power per unit area. Loudness, in sones or loudness units (LU; 1 sone = 975 LU), is the name commonly given to the Ψ intensity. When there are standing waves in the ear canal, the ear canal pressure is a sum of both inward- and outward-traveling pressure waves. It seems reasonable, but has not been adequately proven, that the power flow *into* the ear should be a better measure of hearing performance than the total pressure. Loudness depends in a complex manner on a number of acoustical variables, such as intensity, frequency, and spectral bandwidth, and on the temporal properties of the stimulus, as well as on the mode of listening (in quiet or in noise; binaural or monaural stimulation). Isoloudness contours describe the relation of equal loudness between tones or between narrow bands of noise at different frequencies.

In 1924 Fletcher and Steinberg published an important paper on the measurement of the loudness of speech signals (13). In this paper, when describing the growth of loudness, the authors state

the use of the above formula involved a *summation of the cube root of the energy rather than the energy*.

This cube-root dependence was first described by Fletcher the year before (12). Today any power-law relation between the intensity of the physical stimulus and the psychophysical response is referred to as *Stevens's law* (6,39). Fletcher's 1923 loudness growth equation established the important special case of loudness for Stevens's approximate, but more general, psychological "law."

Cochlear Nonlinearity: How?. What is the source of Fletcher's cube-root loudness growth (i.e., Stevens's law)? Today we know that the basilar membrane motion is nonlinear, and that cochlear outer hair cells (*OHC*s) are the source of the basilar membrane nonlinearity and, as a result, the cube-root loudness growth observed by Fletcher.

From noise trauma experiments on animals and humans, it is now widely accepted that recruitment (abnormal loudness growth) occurs in the cochlea (23). In 1937 Lorente de Nó (22) theorized that recruitment is due to hair cell damage. Animal experiments have confirmed this prediction and have emphasized the importance of OHC loss (40,41). This loss of OHCs causes a loss of the basilar membrane compression first described by Rhode in 1971 (6;42;43, p. 291). It follows that the cube-root loudness growth results from the nonlinear compression of basilar membrane motion due to stimulus-dependent voltage changes within the OHC.

We still do not know precisely what controls the basilar membrane nonlinearity, although we know that it is related to outer hair cell length changes which are controlled by the OHC membrane voltage (43a). This voltage is determined by shearing displacement of the hair cell cilia by the tectorial membrane. We know that the inner hair cell (*IHC*) has a limited dynamic range of less than 60 dB, yet it is experimentally observed that these cells code a dynamic range of about 120 dB (17). Nonlinear compression by cochlear OHCs, prior to IHC detection, increases the dynamic range of the IHC detectors. When the OHCs are damaged, the compression becomes linear, and loudness recruitment results (21).

Loudness Additivity. Fletcher and Munson (2) showed, for tonal stimuli, (a) the relation of loudness to frequency (loudness levels in phons), (b) the dependence of loudness on intensity, (c) a model showing the relation of masking to loudness, and (d) the basic idea behind the critical band (critical ratio).

Rather than thinking directly in terms of loudness growth, they tried to find a formula describing how the loudnesses of several stimuli combine. From loudness experiments with low- and high-pass speech and complex tones (13,15) and from other unpublished experiments over the previous 10 years, they found that loudness adds. Today this model concept is called *loudness additivity*. Their hypothesis was that when two equally loud tones that do not mask each other are presented together, the result is "twice as loud." They showed that N tones that are all equally loud, when played together, are N times louder, for N up to 11, as long as they do not mask each other. Fletcher and Munson found that loudness additivity held for signals between the two ears as well as for signals in the same ear. When the tones masked each other (namely, when their masking patterns overlapped), additivity still held, but over an attenuated set of patterns (2), since the overlap region must not be counted twice. This 1923 model is fundamental to our present understanding of auditory sound processing.

The Argument. Let $G(p_1, p_2)$ be the nonlinear compression function that maps the ear canal pressures p_1 at frequency f_1 and p_2 at f_2 into the loudness in sones, under the condition that the tones are far enough apart in frequency that they do not mask each other. When one tone masks another, the loudness \mathscr{S} is always less than G (i.e., masking always reduces the loudness). When each tone is presented alone, there is no masking, so $\mathscr{S} = G$. It also follows that $\mathscr{S}_1 = G(p_1, 0)$ and $\mathscr{S}_2 = G(0, p_2)$. We assume that G(0, 0) = 0 and $G(p_{ref}, 0) = 1$, where p_{ref} is either 20 μ Pa or the threshold of hearing at 1 kHz. The problem is to find $G(p_1, p_2)$.



Fig. 1. The loudness growth $\mathscr{D}(I)$ at 1 kHz (solid line) from Fletcher (9) in LU (975 LU is 1 sone), along with $2\mathscr{D}(I)$ (dashed line) and $10\mathscr{D}(I)$ (dot-dashed line) for reference. To determine $\alpha *(I)$ draw a horizontal line that crosses the $2\mathscr{D}(I)$ and $\mathscr{D}(I)$ curves, and note the two intensities. The dB difference is $20 \log_{10}(\alpha *(I))$. For example, the total loudness of two 40 dB SL tones presented to each of the two ears is 2000 LU (2 sones), and is equal to the loudness of a single tone at 49 dB SPL. Thus $\alpha *$ at 40 dB is 9 dB.

• Step 1 The pressure p_1 is taken as the reference level for the experiment with $f_1 = 1$ kHz. The level of pressure p_2 , at frequency f_2 , is next determined by requiring that its loudness be equal to that of p_1 . We call this pressure $p^*_2(p_1, f_2)$, since it is a function of both p_1 and f_2 . In terms of the compression function G, p^*_2 is defined by

$$G(0, p_2^*) = G(p_1, 0) \tag{1}$$

• Step 2 Fletcher and Munson scaled the reference pressure p_1 by scale factor α and defined α * such that the loudness of $\alpha * p_1$ is equal to the loudness of p_1 and p^*_2 played together. In terms of *G* this condition is

$$G(\alpha^* p_1, 0) = G(p_1, p_2^*)$$
(2)

This equation defines α *.

Results. For f_1 between 0.8 kHz and 8.0 kHz, and f_2 far enough away from f_1 (above and below) so that there is no masking, 20 log₁₀ $\alpha *(I)$ was found to be ≈ 9 dB for p_1 above 40 dB SL. Below 40 dB SL, this value decreased linearly to about 2 dB for p_1 at 0 phons, as shown in Fig. 1. It was found that the loudness $G(p_1, p^*_2)$ does not depend on $p^*_2(p_1, f_2)$ as f_2 is varied. Thus we may write $\alpha *(p_1, \sqrt{2})$ to show its dependence on p_1 and its independence of p^*_2 .

Fletcher and Munson found an elegant summary of their data. They tested the assumption that

$$G(p_1, p_2) = G(p_1, 0) + G(0, p_2)$$
(3)

namely that the loudnesses of the two tones add. Using Eq. (1), Eq. (3) becomes

$$G(p_1, p_2^*) = 2G(p_1, 0) \tag{4}$$

Combining Eq. (2) and Eq. (4) gives the nonlinear difference equation

$$G(\alpha^*(p_1)p_1, 0) = 2G(p_1, 0)$$
(5)

which determines *G* once $\alpha * (p_1)$ is specified. G(p) may be found by graphical methods or by numerical recursion, as shown in Fig. 136 of Ref. 9 (p. 190).

From this formulation Fletcher and Munson found that at 1 kHz, and above 40 dB SPL, the pure-tone loudness *G* is proportional to the cube root of the signal intensity $[G(p) = (p/p_{ref})^{2/3}$, since $\alpha * = 2^{3/2}$, or 9 dB]. This means that if the pressure is increased by 9 dB, the loudness is doubled. Below 40 dB SPL, loudness was frequently approximated as being proportional to intensity $[G(p) = (p/p_{ref})^2, \alpha * = 2^{1/2}, \text{ or 3 dB}]$. Figure 1 shows the loudness growth curve. Estimated values of $\alpha * (I)$ are given in Fletcher (9, Table 31, p. 192).

The Just-Noticeable Difference in Intensity. Basic to psychophysics, more fundamental than masking, is the concept of a *just-noticeable difference (JND)*, or *difference limen (DL)*. A primary premise of the *auditory theory of signal detection (ATSD)* is that the JND is a measure of the Ψ uncertainty (noise) (6). That is, if we increase the intensity from I to $I + \Delta I$, so that we can just hear the change in intensity, then ΔI should be proportional to the uncertainty of the Ψ representation of I. [ΔI can also have a component due to the signal uncertainty, called "external" noise, as when the signal is "roved," for example (44).] This idea is captured by the equation (Ref. 44, p. 113) $\Delta I = d'\sigma_I$, where ΔI is the intensity JND, and σ^2_I is the variance of I due to the internal noise, *reflected back to the input of the system*. This equation says that the just-noticeable perceptual change in intensity is proportional to the internal noise, as reflected in the intensity variance. This equation defines a model for the detection of the change in signal intensity. While this model is widely accepted, it has not been quantitatively verified (29).

Definition of I(t). The *intensity* of a sound (in watts per square centimeter) is a physical variable, defined as the square of the pressure divided by the acoustic impedance $I = P^2/\rho c$, where P is the root-mean-square (*rms*) pressure, ρ is the density of air, and c is the sound speed. In the time domain when the impedance is fixed, it is common to define the intensity in terms of the normalized time-integrated squared signal pressure p(t), namely

$$I(t) = \frac{1}{\rho cT} \int_{t-T}^{t} p^2(t) dt \tag{6}$$

These two alternative definitions of intensity differ by the integration over and normalization by a fixed interval T seconds long. When the signal p(t) is deterministic, as in the case of pure tones, we shall define T to be the duration of the tone. When the signal is a Gaussian noise, we let $p(t) = \tilde{n}(t)$ [i.e., $N(0, \sigma_n)$], and T be the duration of the noise, leading to $I \equiv \mathscr{F}[\tilde{I}(t)] = \sigma^2_n$.

Definition of ΔI . An *increment in sound intensity* $\Delta I(\alpha)$ may be defined, in terms of a positive pressure scale factor $\alpha \ge 0$ applied to the signal pressure s(t), as

$$I(t, \alpha) = I(t, 0) + \Delta I(t, \alpha)$$
(7)

$$=\frac{1}{\rho cT}\int_{t-T}^{t}[s(t)+\alpha s(t)]^2\,dt \tag{8}$$

Expanding this relation, we have

$$\Delta I(t, \alpha) = (2\alpha + \alpha^2)I(t, 0) \tag{9}$$

The estimate of the intensity $I(t, \alpha)$ is a function of the time and the signal gain, and I(t, 0) indicates the case where $\alpha = 0$.

Definition of an Intensity Detector. In psychophysics the ear is frequently modeled as an intensity detector. It is useful therefore to introduce this popular model formally and compare its performance with that of the ear. We define an *intensity detector* as the intensity I(t) plus the *internal noise* of the detector,

$$\bar{\mathscr{I}}(t) \equiv I(t) + \tilde{\epsilon}(t) \tag{10}$$

which is modeled as an independent Gaussian random process $\tilde{\varepsilon}(t)$, having zero mean and variance σ_{ϵ}^2 [i.e., $N(0, \sigma_{\epsilon})$]. The internal noise limits the *JND* for nonrandom signals such as pure tones.

Throughout this article, we shall only consider zero-mean signals [e.g., s(t)] when using the above definition of the intensity detector. One must carefully distinguish I, the observable intensity, and \tilde{I} , a *decision variable* that is not observable. If we think of the energy detector as a crude model of the ear, \tilde{I} is the decision variable which represents the Ψ intensity (i.e., the loudness). We will show that the intensity detector is *not* a good model of the ear, because both its level dependence and its internal noise are vastly different from those of the ear [i.e., $\mathscr{L}(I) \neq \mathscr{L}(I)$ and $\sigma_{\mathscr{L}}(I) \neq \sigma_{\mathscr{L}}(I)$] (29). However, the intensity detector is an important "straw man" candidate for comparison purposes. It is important to remember that the auditory brain has no access to the intensity of the stimulus. It only receives auditory information from the auditory nerve.

The mean detector intensity is defined by $\mathscr{I} \equiv \mathscr{F}[\tilde{I}]$, and the variance of the detector intensity is defined by $\sigma^2 \mathscr{I}(I, T) \equiv \mathscr{F}[\tilde{I}^2] - \mathscr{I}^2$. From this definition, $\mathscr{I} = I$ and $\Delta \mathscr{I}(\alpha) = \Delta I(\alpha)$. The variance represents the uncertainty of the internal decision variable and plays a fundamental role in the theory of signal detection. We shall see that $\sigma_{\mathscr{I}}$ is a function of both the mean intensity and the duration, but for deterministic signals is simply equal to the internal noise of the energy detector [i.e., $\tilde{I}(t, \alpha)$ is $N(I(\alpha), \sigma_{\epsilon})$]. For stochastic signals $\sigma_{\mathscr{I}}(I, T \to \infty) = \sigma_{\epsilon}$, namely, as the duration of the random signals is increased, $\sigma_{\mathscr{I}}(I, T)$ is finally limited by the internal noise. This means there are conditions (e.g., large T or small I) where the internal noise of the detector will dominate its performance.

Definition of JND_I . The just-noticeable difference in intensity (JND_I) is determined by finding the value of α , which we call α *, such that the subject can correctly identify the decision variable $\tilde{I}(t, \alpha*)$ from $\tilde{I}(t, 0)$ 50% of the time, adjusted for chance. It is convenient and common to refer to JND_I as simply $\Delta I(I)$ rather than using the more cumbersome (but more precise) composite-function notation $\Delta I(\alpha*(I))$.

For the intensity detector defined above, one may analytically determine JND_I and show that $\Delta I(\alpha *) = \sigma_{\epsilon}$. For human subjects we must run an experiment, such as a 2-IFC comparison, and make a model of the observations. In this case the value of α that satisfies the 50%-above-chance discrimination condition, $\alpha *(I)$, depends on I (i.e., $\Delta I/I$ depends on I for tones, but is approximately constant for wideband noise).

Weber's Law. The intensity JND is frequently expressed as a relative JND_I called the Weber fraction, E

defined by $J(\alpha *) \equiv \Delta I/I$. Weber's law J(I), that J is independent of I, was first proposed in 1846 (45).

Weber's Law and Pure Tones. The Weber fraction $J(\alpha *(I))$ is a function of intensity for the most elementary signal, the pure tone (11,16,46). This observation is referred to as the *near miss to Weber's law* (47). The near miss shows that the ear is not an energy detector, since for an energy detector the detector noise σ_{ϵ} is a constant. For recent discussions of why Weber's law holds approximately for tones (48;49, p. 721), or why it holds for wideband noise more than 20 dB above threshold, we refer the reader to the helpful and detailed reviews by Viemeister (50), Hartmann (51), and Allen and Neely (29).

The JND for an Energy Detector. Next we review the derivation of the JND for the energy detector. Two independent signals [either s(t) or $\tilde{n}(t)$, $N(0, \sigma_n)$] are presented to the energy detector with gains $\alpha = 0$ and α , having the decision variable $\tilde{I}(t, \alpha)$. We would like find α such that the more intense signal is greater than the less intense signal 75% of the time. This task is equivalent to the following: Find α such that the difference

 $\tilde{\Delta I}(t, \alpha) \equiv \tilde{I}(t, \alpha) - \tilde{I}(t, 0)$ is greater than zero 75% of the time. It is assumed that $\tilde{\Delta I}$ is Gaussian with a variance of $2\sigma^2_{\mathcal{I}}$ [the variances $\sigma^2_{\mathcal{I}}(0)$ and $\sigma^2_{\mathcal{I}}(\alpha)$ are assumed to be similar enough to be treated as equal].

When $\Delta \mathscr{I}(\alpha) \equiv \mathscr{C}[\Delta I(t, \alpha)]$ is $\sigma_{\mathscr{I}}$ [i.e., when $\Delta I(t, \alpha)$ is $N(\sigma_{\mathscr{I}}\sqrt{2} - \sigma_{\mathscr{I}})$], the probability that $\Delta \mathscr{I} > 0$ is ≈ 0.76 . This probability is close enough to the definition of 0.75 that it has been adopted as the de facto standard detection threshold (44). The ratio of $\Delta \mathscr{I}$ to $\sigma_{\mathscr{I}}$ is an important statistic of the experimental signal uncertainty called d':

$$d'(\alpha) \equiv \frac{\Delta \mathscr{I}(\alpha)}{\sigma_{\mathscr{I}}(\alpha)} \tag{11}$$

Using this notation, the definition of $\alpha *$ is $d'(\alpha *) \equiv 1$.

Thus with the assumption of an intensity detector having Gaussian detection variables of equal variance, and a detection criterion of 76%, the Weber fraction is

$$J = \frac{\sigma_{\mathscr{I}}(I)}{\mathscr{I}} \tag{12}$$

This expression follows from the definition of J and the fact that $\mathscr{I} = I$ and $\sigma_I = \sigma_{\mathscr{I}}$. The ratio of the intensity to the intensity variance defines an intensity signal-to-noise ratio $\text{SNR}_{\mathscr{I}} \equiv \mathscr{I}/\sigma_{\mathscr{I}}$ which is more intuitive than the Weber fraction.

The gain α * is then determined from Eq. (9) and Eq. (12):

$$2\alpha^* + (\alpha^*)^2 = \frac{1}{\mathrm{SNR}_{\mathscr{S}}} \tag{13}$$

and since $\alpha * \ge 0$, we have $\alpha * = \sqrt{1 + 1/\text{SNR}_{\mathcal{I}}} - 1$. This last equation allows one to calculate $\alpha *$ given SNR₃, or estimate SNR₃ given a measurement of $\alpha *$ from a pure tone intensity *JND* experiment.

Internal versus External Noise. There are two commonly identified types of masking noise, internal and external noise. Both of these two types of noise are modeled as Ψ (e.g., loudness) variability, which is synonymous with masking. Internal noise is due to the stochastic nature of the Ψ representation, while external noise is due to the stochastic nature of the Ψ representation, while external noise is due to the stochastic nature of the stimulus), which is transformed into Ψ variability by the auditory system. Internal noise sets the fundamental limit on the representation of the auditory system. Roving the signal is a technique designed to make the external noise dominate.

For the case of an external noise, it is possible to show (M. M. Sandhi, personal communication) that

$$\sigma_{\mathcal{I}}^{2}(I,T) = \frac{2I^{2}}{T^{2}B^{2}} + \sigma_{\epsilon}^{2}$$
(14)

where B is an effective bandwidth (that depends on T) defined by

$$B(T) = \left(\int_{\tau=0}^{T} \int_{t=0}^{T} \rho^{2}(t-\tau) \, dt \, d\tau\right)^{-1/2}$$

and $\rho(t - \tau) \equiv \mathscr{C}[\tilde{n}(t)\tilde{n}(\tau)]/I$ is the normalized [i.e., $\rho(0) = 1$] covariance of the stochastic signal $s(t) = \tilde{n}(t)$. Thus for the intensity detector with a Gaussian input having a variance that dominates the detector noise, Weber's

law holds, and $J(\sqrt{\rho_e^2(t) + 1/\text{SNR}_I}) = \sqrt{2}/TB$, or $\text{SNR}_{\mathcal{F}} = TB/\mathcal{L}(I) = \mathscr{E}[\mathcal{L}(I)]$ The product TB is called the degree-of-freedom parameter.

Definition of $\Delta \mathscr{G}$. Any superthreshold increment in the sound intensity must have a corresponding loudness increment. A *loudness increment* $\Delta \mathscr{G}(I)$ is defined as the change in loudness $\mathscr{G}(I)$ corresponding to an intensity increment $\Delta I(I)$. When $\Delta I(I)$ is JND_{*I*}, the corresponding $\Delta \mathscr{G}$ defines the *loudness just-noticeable difference* JND_{*G*}. Just as ΔI is commonly used to describe JND_{*I*}, we shall use $\Delta \mathscr{G}$ to describe JND_{*G*}.

While it is not possible to measure $\Delta \mathscr{D}$ (i.e., JND_{\mathscr{D}}) directly, we assume that we may expand the loudness function in a Taylor series, giving

$$\mathcal{L}(I + \Delta I) = \mathcal{L}(I) + \Delta I \left. \frac{d\mathcal{L}}{dI} \right|_{I} + \text{HOT}$$

where HOT represents higher-order terms that we shall ignore (29). While it is not meaningful to form a Taylor series expansion of the single trial loudness $\tilde{l}(t, I)$, it is meaningful to expand the expected value of this random process. If we solve for $\Delta \mathscr{G}$, defined as

$$\Delta \mathcal{L} \equiv \mathcal{L}(I + \Delta I) - \mathcal{L}(I) \qquad (15)$$

we find

$$\Delta \mathcal{L} = \Delta I \left. \frac{d\mathcal{L}}{dI} \right|_{I} \tag{16}$$

We call this expression the *small-intensity-increment* approximation. It shows that the loudness JND $\Delta \mathscr{G}(I)$ is related to the intensity JND $\Delta I(I)$ by the slope of the loudness function evaluated at intensity *I*.

From the Taylor expansion the internal loudness standard deviation may be related to an external effective intensity variance by

$$\sigma_{\mathcal{L}} = \sigma_I \left| \frac{d\mathcal{L}}{dI} \right|$$

It follows that $d' = \Delta \mathscr{D}/\sigma_{\mathscr{D}}$ and that JND $_{\mathscr{D}}$ is defined by d' = 1 in a manner identical to the definition of JND_{*I*}.

The Loudness SNR. We define the loudness SNR as $SNR_{\mathscr{D}} \equiv \mathscr{D}/\sigma_{\mathscr{D}}$. From the definitions of $SNR_{\mathscr{D}}, d'$, and J,

$$\mathrm{SNR}_{\mathcal{L}} = d' \left(J(I) \frac{d \mathcal{L}_{\log}}{d\beta} \right)^{-1} \tag{17}$$

where $\beta \equiv 10 \log_{10}(I/I_{\text{ref}})$ and $\mathscr{D}_{\log}(\beta) \equiv 10 \log_{10} \mathscr{D}(10^{\beta/10})$. This equation is important because (1) all the terms are dimensionless, (2) we are used to thinking of the loudness and intensity on a log scale (as in Fig. 1), and (3) $d\mathscr{D}_{\log}/d\beta$ is constant at large intensities (because, according to Stevens's law, $\mathscr{D}(I)$ is a power law). To estimate the power-law slope using $\gamma \equiv d\mathscr{D}_{\log}/d\beta$ it is necessary to treat \mathscr{D} as an intensity when defining \mathscr{D}_{\log} .

A much simpler way to write Eq. (17) is to define $\text{SNR}_I \equiv I/\sigma_I$, which along with $\Delta I = d'\sigma_I$ gives (29)

$$SNR_I = \gamma SNR_{\mathcal{A}}$$
(18)

This equation says that the loudness SNR and the intensity SNR are related by the exponent γ of the loudness power-law function.

Masking. Masking is the elevation in threshold due to a masking signal. To define masking we must first define the *masked threshold*. The energy of the masker spreads out along the basilar membrane with a density given by $I_x(f_m, I_m, x)$, where x(f) is the *characteristic place* corresponding to frequency f. To model the masked threshold we need a model of $I_x(f_m, I_m, x)$ near the probe place $x(f_p)$.

The Masked Threshold. The hearing threshold in the presents of a masking signal is called the masked threshold. Since it is used in the definition of masking, it is a more fundamental than masking. More formally, the masked threshold $I^*{}_p(f_p, I_m)$ is the threshold intensity of a probe (maskee) $I^*{}_p$ at frequency f_p in the presence of a masking signal having intensity I_m . When the masker intensity is set equal to zero, the masked threshold is just the probe intensity at the threshold of hearing in quiet, or the unmasked threshold $I^*{}_p(f_p) \equiv I_{*p}(f_p, I_m = 0)$. As before, the asterisk indicates that special value of I_p which gives a 75% correct score for the detection of the probe in a 2-IFC task, due to the loudness uncertainty characterized by σ_{s} .

Because the hearing threshold is generally defined statistically as the probe intensity corresponding to the 50% correct score (corrected for chance) for detecting the probe from some Ψ decision random variable, the masked threshold is not a random variable. To model masking we must first identify the Ψ decision random variable and then model the masked threshold $I^*_p(f_p, I_m)$.

Masking and the Masking Pattern. The masking M is defined as the ratio of the masked to the unmasked threshold:

$$M(f_{\rm p}, I_{\rm m}) \equiv \frac{I_{\rm p}^*(f_{\rm p}, I_{\rm m})}{I_{\rm p}^*(f_{\rm p})}$$

The masked threshold is frequently reported in dB SL (i.e., as $10 \log M$), where SL means that the masked threshold is referred to the *sensation level* (i.e., the unmasked threshold). The *masking pattern* is a description of the masking as a function of the masker level and the probe frequency. The masker can be any signal, such as a tone, narrowband noise, wideband noise, or even speech.

The masked threshold $I^*_{\rm p}(f_{\rm p}, I_{\rm m})$ is frequently measured with a pure-tone probe signal; however, a narrow probe band of noise centered on frequency $f_{\rm p}$ is sometimes used to reduce the beating that can take place when the masker is a pure tone. In this case it seems logical to measure the unmasked threshold $I^*_{\rm p}(f_{\rm p})$ with the same probe signal when computing the masking.

Definition of $\Delta \mathscr{I}$ for Masking. We repeat the derivation of the intensity detector developed for JND, but this time using a probe that differs from the masker. As in the derivation of Eq. (8), an increment in the intensity detector output $\Delta \mathscr{I}(\alpha)$ is defined in tems of a pressure scale factor α applied to the probe signal $\tilde{p}(t)$:

$$\tilde{\mathscr{I}}(t,\alpha) = \tilde{\mathscr{I}}(t,0) + \widetilde{\Delta\mathscr{I}}(t,\alpha)$$
(19)

$$= \frac{1}{\rho cT} \int_{t-T}^{t} [\tilde{n}(t) + \alpha \tilde{p}(t)]^2 dt + \tilde{\epsilon}$$
 (20)

where $\tilde{n}(t)$ is the masker and $\tilde{p}(t)$ is the tone probe (maskee). As a natural generalization of Eq. (8), we set the intensity of the probe equal to that of the masker (i.e., $I = \sigma_p^2 = \sigma_n^2$), and control the intensity of the probe

with the scale factor α . Expanding Eq. (20) and taking the expected value gives

$$\frac{\Delta \mathscr{I}(t,\alpha)}{\mathscr{I}} = \frac{2\alpha}{\rho cT} \int_{t-T}^{t} \rho_{np}(t) \, dt + \alpha^2 \tag{21}$$

where

$$\rho_{np}(t) \equiv \epsilon [\tilde{n}(t)\tilde{p}(t)]/I$$
(22)

is the normalized correlation coefficient between the masker and probe. When ρ_{np} is nonstationary, it is a function of time *t*, and when it is stationary, it is constant over time and can come out of the integral, which then integrates to 1. To simplify the notation, we define the effective correlation $\rho_{e}(t)$ as the integral of $\rho_{np}(t)$ over the *T*-second rectangular window,

$$\rho_{\mathsf{e}}(t) \equiv \frac{1}{\rho cT} \int_{t-T}^{t} \rho_{np}(t) \, dt \tag{23}$$

Equation (21) defines the relative size of the intensity detector's output $\Delta \mathscr{I}/\mathscr{I}$ as a function of α . If we require that $\Delta \mathscr{I}$ be at the detection threshold relative to the magnitude of the detector's internal noise $\tilde{\varepsilon}$, then we may solve for α *. In terms of the defacto detection measure d' [Eq. (11)],

$$d'(\alpha) \equiv [2\alpha \rho_{e}(t) + \alpha^{2}] \text{SNR}_{I}$$

Since $d'(\alpha *) = 1$ defines $\alpha *$,

$$2\alpha^* \rho_{\mathbf{e}}(t) + (\alpha^*)^2 = 1/\mathrm{SNR}_{\mathscr{I}}$$
(24)

Because $\alpha \ge 0$ by definition, the solution to this equation is $\alpha * = \sqrt{\rho_{c}^{2}(t) + 1/\text{SNR}_{I}} - \rho_{e}(t)$. The correlation between $\tilde{n}(t)$ and $\tilde{p}(t)$ is bounded between $-1 \le \rho_{np}(t) \le 1$; thus

$$\sqrt{\rho_{\mathsf{e}}^2 + \frac{1}{\mathrm{SNR}_{\mathscr{I}}}} - |\rho_{\mathsf{e}}| \le \alpha^* \le \sqrt{\rho_{\mathsf{e}}^2 + \frac{1}{\mathrm{SNR}_{\mathscr{I}}}} + |\rho_{\mathsf{e}}|$$

This inequality bounds the range of $\alpha * (\rho_e, SNR_{\mathcal{J}})$ for the energy detector, for the case of d' = 1.

Classes of Masking. The most basic classes of masking are simultaneous and nonsimultaneous masking. In this article we only consider simultaneous masking.

Frozen versus Random Maskers. Noise maskers come in two important forms: so-called frozen-noise and random-noise maskers. The term *frozen noise* is an oxymoron because the word *noise* is synonymous with stochastic, whereas *frozen* is synonymous with deterministic. We shall call such signals *high-degree-of-freedom signals*, or simply *frozen signals*, but never "frozen noise." Live music is an example of a stochastic signal, whereas recorded music is an example of a high-degree-of-freedom signal.

As described by Eq. (14), the variance of a random masker can increase the masking. This effect has been called *external noise*. It is important to determine the relative contribution of the variance of the stimulus and the internal representation. This may be done by measuring the masked threshold twice: once with the random masker, and again with it frozen. If the two masked thresholds are the same, the internal noise is greater than the external noise.

Wideband Maskers. When the masking signal has a wide bandwidth, the energy is spread out along the basilar membrane. For wideband signals, the degree of this correlation across frequency can be important in reducing the external noise. Because of the filtering and the nonlinear properties of the cochlea, it is necessary to understand narrowband masking before we attempt to analyze wideband maskers.

Narrowband Maskers. When the masking signal has a narrow bandwidth, the spread of the energy along the basilar membrane is limited by the filtering properties of the cochlea. When the signal is deterministic or of long duration, the JND is limited by the internal noise.

There are two basic classes of narrowband masking measurements, called masking patterns (MPs) and psychophysical tuning curves (PTCs). The masking pattern is specified in terms of a fixed masked and a variable probe, while the *psychophysical tuning curve* is specified in terms of a fixed low-level (i.e., near-threshold) probe and a variable masker. Because of the nonlinear compressive properties of the cochlea, the difference between the MP and PTC, which is quite large, is important, as it gives insight into the nonlinear properties of the cochlea. We shall only deal with the MP here.

There are three basic regions of a masking pattern, corresponding to the *downward spread of masking* $(f_p < f_m)$, the *upward spread of masking* $(f_p > f_m)$, and *critical-band masking* $(f_p \approx f_m)$. Critical-band masking is the realm of several poorly understood, but important, masking issues, including the *linearity of masking* (an extension of Weber's law to the case of masking), the *asymmetry of masking* (the dB difference in masking between a tone and a narrow band of noise of equal intensity), and *beats*. When the frequency difference between the masker and the probe (maskee) becomes greater than the cochlear filter bandwidth, the masking depends on the shape of the cochlear filters and the cochlear nonlinear compression, which determine the properties of the upward and downward spread of masking.

Critical-Band Masking and Beats. Beating occurs when the masker and probe signals are correlated, as when two or more tones are within the bandwidth of a single cochlear filter (e.g., critical-band masking). This was the case for Riesz's 1928 experiment where the probe and masker were tones separated by 0.2 Hz to 35 Hz (16). The presence of beats is quantified for the energy detector by $\rho_e(t)$. Within the cochlear filter bandwidth (i.e., the critical band) the signal pressure components add in a linear manner. It is frequently said that the power of the components adds, but this is incorrect; power adds only when $\rho_e = 0$, namely when there are no beats. Even though beat detection only occurs in a small frequency region around the critical band where the signal envelopes are correlated, it is critical to understand it quantitatively.

As the tones are moved apart in frequency, the signal develops a maximum roughness quality when the cochlear filter bandwidth is reached. This shows that the temporal integration has a bandwidth that is greater than the critical bandwidth. When the frequency difference is greater than a critical band, the signals become independent ($\rho_e = 0$), the beating disappears, and the loudnesses of the masker and probe, presented together, add in magnitude, resulting in a total loudness that is always greater than the loudness of the masker alone.

Modulation Detection. As may be seen from Eq. (21), $\Delta \mathscr{I}$ depends on two terms, a correlation term $\alpha \rho_{\rm e}(t)$ and a fixed term α^2 . When $\max |\rho_{\rm e}(t)| > \alpha$, the correlation term dominates, and when $\max |\rho_{\rm e}(t)| < \alpha$ the quadric term dominates. Thus when $f_{\rm m} \approx f_{\rm p}$, ΔI is time-varying around zero, and the de facto formula (11) with d' = 1, derived under the assumption that $\Delta I > 0$, fails. When the mean loudness does not change (i.e., critical band masking), the CNS must use a different criterion, which is characterized by the ratio of the variances, as described by a maximum likelihood analysis (54). This critical-band detection paradigm is called *intensity modulation detection* (MD_I) to reflect the idea that the mean intensity is zero (or close to zero). Riesz's experiment, which is a special case of narrowband masking signals, is a classic example of MD_I (50).

The Masked Threshold as an Internal Noise. In 1947 Miller pointed out that Riesz's "JND" experiment is formally a masking task, since the probe and the masker are not the same signal. He then demonstrated a close similarity between Riesz's modulation detection threshold and the wideband-noise JND_I. But practically speaking, Riesz's experiment is measuring a form of JND, because the masker and maskee are close in frequency (viz., 3 Hz apart). The term "close in frequency" is not well defined, but is related to the region of beats. The maximum rate of loudness variation is believed to be limited by an internal *temporal integrator* having a time

constant between 100 ms and 300 ms, corresponding to a low-pass filter having a 3 dB bandwidth between 0.5 Hz and 1.6 Hz (Ref. 6, section on "Temporal Integration"). For example, Munson (30) found a single-pole integrator with an integration time constant of 200 ms, and Riesz showed that 3 Hz is the optimum frequency of modulation for his detection task. However, the 3 dB bandwidth is not a meaningful characterization of the perceptual bandwidth, since experiments show we can hear beats up to at least 20 Hz or 30 Hz. This implies that the integrator should have a shallow slope and that perhaps the perceptual bandwidth should be specified at the -30 dB point rather than the -3 dB point, assuming Munson's single-pole filter. In summary, Riesz and MIIIer provided us with the insight that the masked threshold may be modeled in terms of the same (internal) noise that limits the JND (8,29).

It follows that the loudness uncertainty $\sigma_{\mathscr{D}}$ for Miller's wideband noise is similar in magnitude to $\sigma_{\mathscr{D}}$ for pure tones as measured by Riesz. Following Miller's lead, we define the masked threshold in terms of the Ψ uncertainty (e.g., "loudness noise"). If $\tilde{l}_m = \tilde{l}(I_m)$ is the single-trial loudness due to a masker at intensity I_m , and $\tilde{l}(I_m, I_p)$ is the single-trial loudness due to both the masker and a probe of intensity I_p , then $\Delta l(I_m, I_p, \rho_e(t)) = \tilde{l}(I_m, I_p) - \tilde{l}_m$ defines a decision variable for the masked threshold. The effective correlation $\rho_e(t)$ is used to account for the correlations between probe and masker corresponding to the critical-band region and beats. The probe intensity at the masked threshold $I^*_p(f_p, I_m)$ is defined as that value of I_p such that the probability of detecting the probe is 50% correct, corrected for chance.

The Loudness Signal-To-Noise Ratio

We have interpreted the pure-tone JND as a measure of the Ψ noise. In this section we complete this interpretation by calculating the loudness SNR required to account for the pure-tone and wideband-noise JND_I. In the following we directly compare the tonal loudness growth function $\mathscr{D}(I)$ of Fletcher and Munson (2) measured by Munson in 1932 (55) with the tonal intensity JNDs $\Delta I(I)$ from Riesz (16). Both sets of experimental data were taken in the same laboratory within a few years of each other, and it is likely they used the same methods and the same equipment, given its cost. This will allow us to estimate the loudness JND $\Delta \mathscr{D}(\mathscr{D})$, and therefore the *loudness signal-to-noise ratio* (SNR $_{\mathscr{D}} \equiv \mathscr{D}(\sigma_{\mathscr{D}})$. JND data are quite sensitive to the experimental measurement conditions (20, pp. 141–143). The Riesz (16) and Munson (55) data are interesting because they are taken under conditions similar to the loudness data of Fletcher and Munson, which were continuous (1 s long) pure tones.

A Direct Estimate of $JND_{\mathscr{D}}$. In Fig. 2 we present a direct estimate of the loudness $JND_{\mathscr{D}}[\Delta \mathscr{D}(\mathscr{D})]$ computed from Eq. (16) at all 11 frequencies Fletcher and Munson used to measure the loudness. The procedure for doing this is described in Allen and Neely (29). Each of the four displays a different frequency range. As indicated in the figure legend, we have marked the point on the curve where the slope changes. For the 62 Hz data in the upper left panel we see that $\Delta \mathscr{D}$ is constant for levels below about 50 dB SL. Over most of the frequency range, below 20 dB [$\mathscr{D}(I) < 100$ LU], we find $\Delta \mathscr{D} \propto \sqrt{\mathcal{L}}$ Between 20 dB and 60 dB [$100 < \mathscr{D}(I) < 3000$], we find $\Delta \mathscr{D} \propto \mathscr{D}^{1/3}$. Above 60 dB [$\mathscr{D}(I) > 3000$], we find $\Delta \mathscr{D} \propto L$. Thus the loudness and $JND_{\mathscr{D}}$ are proportional above 60 dB SL.

In Fig. 2 on the lower left we also show $\Delta \mathscr{D}(\mathscr{D})$ for Miller's (4) wideband-noise JND_I data. Miller gives the loudness level as well as the intensity JND measurement. We converted this loudness level to loudness using Fletcher and Munson's (2) reference curve (i.e., Fig. 1). The SNR $_{\mathscr{D}}$ for the tones and the wideband noise are almost identical, especially over the frequency region between 0.25 kHz and 8.0 kHz.

Determination of the Loudness SNR. Given the $d' \equiv \Delta \mathscr{D}/\sigma_{\mathscr{D}}$ and $\text{SNR}_{\mathscr{D}} \equiv \mathscr{D}/\sigma_{\mathscr{D}}$, it follows that $\mathscr{D}/\Delta \mathscr{D} = \text{SNR}_{\mathscr{D}}/d'$. From Fig. 3 for levels above 65 dB, the $\text{SNR}_{\mathscr{D}}$ becomes constant. From Fig. 2, $\text{SNR}_{\mathscr{D}}$ increases by a factor of 2 when the loudness increases by a factor of 4, up to about 55 dB.

As an application of Eq. (18), we calculate $SNR_{\mathscr{D}}$ for Miller's wideband masking data. Miller found J = 0.1, which gives $SNR_I = d'/J = 10$. As shown in Fig. 4 on the upper right, the power law has a slope of



Fig. 2. $\Delta \mathscr{G}(\mathscr{G}, f)$ computed directly from Eq. (16) using Riesz's JND data and the Fletcher–Munson loudness–intensity curve, for levels between 0 and 120 dB SL. The mark \odot has been placed on the curves at an intensity of 55 dB SL for 62 Hz and 125 Hz, 60 dB SL for 0.25 kHz to 1 kHz, 55 dB SL for 2 kHz to 5.65 kHz, and 50 dB SL for 8 kHz to 16 kHz. In the upper right panel we have added straight lines for reference, having slopes of $\frac{1}{2}$, $\frac{1}{3}$, and 1, for levels between 0 and 20 dB SL, between 20 dB SL and 60 dB SL, and above 60 dB SL, respectively. From these plots it is clear that $\Delta \mathscr{G}(\mathscr{G})$ is described by a power law in \mathscr{G} having three straight line segments. Between 0 and 20 dB SL, the slope is close to 0.5. Between 20 dB SL and 60 dB SL the slope is close to $\frac{1}{3}$ ($\Delta \mathscr{G} \propto \mathscr{G}^{1/3}$). Above 60 dB SL, the slope is 1 ($\Delta \mathscr{G} \propto \mathscr{G}$). Fechner's hypothesis [$\Delta \mathscr{G}(\sqrt{\mathcal{L}})$] appears to hold only for 62 Hz and 125 Hz below 50 dB SL. One extra curve, labeled with the symbol +, has been added to the lower left panel, showing $\Delta \mathscr{G}(\mathscr{G})$ for the wideband noise case of Miller (4). This curve has a slope of approximately 0.63 for \mathscr{G} less than 10³, and then merges with the tone data up to a loudness of 10⁵, the upper limit of Miller's data.

 $\gamma = \frac{1}{4}$ above 40 dB SL. Thus SNR₂ \approx 40, and this estimate is in fair agreement with estimates for pure tones as shown in Fig. 3.

Summary of $JND_{\mathscr{D}}$ Results. The pure-tone and wideband-noise JND results have been summarized in terms of $SNR_{\mathscr{A}}(\mathscr{D})$. These curves seem similar enough that they may be characterized by one curve, at least for coding purposes. Between threshold and 60 dB SL, $\sigma_{\mathscr{D}} \propto \mathscr{D}^p$ with $\frac{1}{3} \leq p \leq \frac{1}{2}$ for tones and p = 0.63 for noise; above 60 dB SL, $\sigma_{\mathscr{D}} \propto \mathscr{D}$. Thus it appears that once we know the signal intensity, we know the loudness SNR for any signal bandwidth. Next we will look at masking data and describe how to use this information.

Narrowband Maskers

Narrowband maskers, which result in a limited region of energy spread along the basilar membrane, hold the key to understanding masking. The two maskers we shall consider are a pure tone and a subcritical band of noise.



Fig. 3. Plot of $\mathscr{D}(I)/\Delta \mathscr{D} = \text{SNR}/d'$ against loudness for pure tones having intensities between 0 and 120 dB SL. Each of the four panels is for a different frequency range. Except at 62 Hz and 125 Hz, and below about 60 dB SL, the loudness signal-to-noise ratio SNR/(I) is proportional to \mathscr{D}^{1-p} , where $\frac{1}{2} \le p \le \frac{1}{3}$. Above 60 dB SL the SNR/(I) saturates with a value between 30 and 50 linear units. For 62 Hz and 125 Hz the SNR/(I) for the wideband noise signal of Miller (4). Even though the conditions are very different, it the SNR/(I) is similar for tones and wideband noise.

Tone-on-Tone Masking. When one tone is used to mask another tone, band-pass masking patterns result, as shown by the dashed curve of Fig. 5 (5), corresponding to an $f_{\rm m} = 400$ Hz tonal masker at 65 dB SL ($I_m = 80$ dB SPL), as a function of probe frequency $f_{\rm p}$. Such patterns, as published by Fletcher (12) and Wegel and Lane (1), were used by Fletcher and Munson to derive the theory of loudness (2,37). An alternative way to view these data is as masking level curves as shown in Fig. 6 for a $f_{\rm m} = 400$ Hz masker. Such data were first published in Fletcher (12,56) and Wegel and Lane (1) for frequencies $f_{\rm p}$ between 0.25 kHz and 4 kHz and intensities between 0 and 85 dB SL. In 1922 Fletcher and Wegel published a major study (10) that accurately measured the threshold of hearing for the first time (37, p. A7), and the masked threshold measurements of Fig. 6 followed from this 1922 experiment.



Fig. 4. In 1947 Miller (4) measured JND_{*I*} and the loudness level for two subjects using wideband noise (0.15 kHz to 7 kHz) for levels between 3 dB SL and 100 dB SL. The intensity of the noise was modulated with a ramped square wave that was high for 1.5 s and low for 4.5 s. The loudness, computed from Miller's phon data using Fletcher and Munson's (2) 1-kHz-tone loudness growth curve (solid curve) is shown in the upper left panel, along with the Fletcher–Munson tonal loudness growth function (dashed curve). The upper right panel shows the slope $\gamma(I) \equiv d \mathscr{L}_{\log}/d\beta$ of Miller's loudness growth function (two subjects). For reasons that are not clear, the loudness values at the limits seem to be in error. In the lower left panel we plot (two subjects) $\log_{10} \Delta \mathscr{L}(\mathscr{D})$ as a function of $\log_{10} \mathscr{D}$. This curve has a slope that is close to 0.63 for $1 < \Delta \mathscr{L} < 100$ (i.e., $\Delta \mathscr{D} \approx \mathscr{L}^{0.63}$). In the bottom right panel we show the average SNR $_{\mathscr{D}}$ for the two subjects. As in the upper right panel, the ends of the range seem to be in error.

It is interesting to compare the 1923 Fletcher data (12) of Fig. 6 with the 1950 data of Fig. 5. For example, the masked threshold for a 65 dB SL masker is shown in Fig. 5 with \circ symbols. The largest difference is about 17 dB at 2 kHz. A careful comparison between the two data sets shows that a 58 dB SL masker at 400 Hz from the 1923 Fletcher data (12) (shown in Fig. 5 as + symbols) is within a few decibels of the 65 dB SL masker for Egan and Hake's data (5), for all probe frequencies. Subject variability is one obvious explanation for these differences. An alternative is that Fletcher, Wegel, and Lane may have compensated for the $2f_1 - f_2$ distortion product in their measurements, as they did for the subjective harmonics.



Fig. 5. The solid curve is the simultaneous masking pattern for a 90 Hz band of noise centered at 410 Hz having an rms intensity of 65 dB SL (a spectral level of 45.6 dB, as shown by the short solid line). The dashed curve is the masking pattern for a 400 Hz pure tone at 65 dB SL (circle). The probe signal (maskee) was a 0.7 s pure tone. Note the large (26 dB) difference in the masked threshold at the masking frequency of 410 Hz. When the probe is near 500 Hz, the distortion product $2f_1 - f_2$ is the limiting factor in detection. The tone masking curve seems to be shifted to higher frequencies by a ratio of about 1.2 (a ratio of 600 Hz and 500 Hz). The dips at 0.8 kHz and 1.2 kHz are due to subjective harmonics (9). The masked threshold for a 400 Hz tone as determined from the data of Fig. 6 of 12 is shown by symbols + for 58 dB SL and \circ for 65 dB SL.

The Spread of Masking. As may be seen from Fig. 6 (solid lines), for the case $f_p = 2$ kHz to 4 kHz, the onset of masking is abrupt at about 60 dB to 65 dB and has a slope of about 2.4; namely,

$$\frac{I_{\rm p}^{*}(f_{\rm p}, I_{\rm m})}{I_{\rm ref}} = \left(10^{-6} \frac{I_{\rm m}}{I_{\rm ref}}\right)^{2.4} \tag{25}$$

This expression is plotted as the short-dashed line superimposed on the 3 kHz curve of Fig. 6. This steep slope is referred to as the *upward spread of masking*. For *downward spread of masking* ($f_p < f_m$), the growth of masking is a compressive power law in intensity (dashed lines).

Critical-Band Masking. For probe frequencies near the masker frequency of 400 Hz the masking is said to be *linear in intensity*. For example, at $f_p = 0.45$ kHz (dash-dot line in Fig. 6) the masking curve is well approximated by the linear relation

$$\frac{I_{\rm p}^*(f_{\rm p}, I_{\rm m})}{I_{\rm m}} = \frac{1}{40} \tag{26}$$

for $I_{\rm m}$ greater than about 25 dB SL, as indicated by the dotted line superimposed on the 0.45 Hz masking curve (1, p. 270). Other examples of this linearity include the masking of tones by narrow bands of noise (2,5,31) and the masking of tones and narrow bands of noise by wideband noise (3,9,25,31,57).



Fig. 6. Tone-masking-tone data from Fletcher (12) and Wegel and Lane (1) for a masker at 400 Hz. The dashed lines correspond to probe frequencies between 0.25 kHz and 0.45 kHz, while the solid lines correspond to probe frequencies of 1 kHz to 4 kHz. The masking at 0.45 kHz is proportional to the masker level (i.e., the slope is close to 1). For 2 kHz, 3 kHz, and 4 kHz there is a threshold effect at about 60 dB SL. For these frequencies the slope is greater than 1. The short dashed line superimposed on the 3 kHz curve is given by Eq. (25) and has a slope of 2.4. This steep slope is an important characteristic of the *upward spiral of masking*.

Equation (26) is an extension of Weber's law for JNDs to the case of masking. It is just as important to understand (i.e., model) the linearity of masking as it is to understand why Weber's JND law holds for wideband noise, as the explanations are the same.

While the linearity of masking seems to be a trivial experimental observation, it is a surprising result. When the probe is added to the masker and the two signals are within a critical bandwidth, their basilar membrane motion adds (e.g., two sine waves beat). However, the response level of the basilar membrane motion, the neural response, and the resulting loudness are all nonlinear functions of level. Thus it is not initially obvious why the masking should be proportional to the intensity.

Linearity of Masking and Weber's Law. If J is approximately constant for $f_p = f_m$, then it is reasonable to expect that it will be approximately constant when $f_p \approx f_m$. If we interpret I_p as the change in intensity due to the probe, then $\Delta I = I^*_p$ and $I^*_p/I_m \equiv \Delta I/I$. Thus Eq. (26) is an extension of Weber's JND law to masking, but is *not* Weber's law, becuse that law strictly applies to the JND. Clearly however the two cases are functionally equivalent. Riesz was the first to recognize this important correspondence. Five years after Fletcher published the masking level curves, Riesz (16) executed an extensive quantitative study of the critical-band region.

Riesz came to two important conclusions. First, he interpreted ΔI in terms of a short-term intensity variation, and defined $\Delta I = I_{\text{max}} - I_{\text{min}}$. With his interpretation of ΔI as a short-term intensity, he was able to precisely test Weber's law under conditions of masking. Thus Riesz's experiment was the first to make the important connection between critical-band masking and Weber's law.

Second, unlike Wegel and Lane's conclusion that I_p^*/I_m is a constant (i.e., that Weber's law holds), Riesz found that $\Delta I/I$ is not exactly constant. In other words, upon careful scrutiny, he showed that Eq. (26) does not hold exactly for the case of tones. Unfortunately it was almost 20 years before Riesz's observations were fully appreciated (4,8).

Riesz's JND Experiment. According to Eq. (21), $\Delta \mathscr{I} = 0$ when the temporal integration time *T* is long relative to the time variations of $\rho_{np}(t)$ [Eq. (22)]. It is difficult to argue that ΔI is proportional to *I* (i.e., Weber's law) if $\Delta I = 0$.

Riesz found a trivial resolution of this problem. He assumed that the ear averages over a short enough interval that it can track the variations over time. This idea is obvious, because one can hear the slow beating of two sine waves as their loudness slowly varies. From this point of view, Riesz defined his measure of JND_I as

$$\frac{\Delta I}{I} = \frac{(1+\alpha^*)^2 - (1-\alpha^*)^2}{(1-\alpha^*)^2} \tag{27}$$

which is $(I_{\text{max}} - I_{\text{min}})/I_{\text{min}}$. For small α , Riesz's formula reduces to $J \approx 4\alpha$, which is similar to the first right-hand term in Eq. (21). When Riesz reports J = 0.1, we have $\alpha * = 0.025$.

If Riesz had ignored the beating and treated the two tones as independent, then ΔI would have been the intensity of the two tones played together minus the intensity of the masking tone alone, and he would have reported the Weber fraction as $J_i = [(1 + \alpha^2) - 1]/1 = \alpha^2$, which is the second term in Eq. (21). Thus given his actual measure value of $\alpha * = 0.025$, he would have reported $J_i = 0.000625$ rather than 0.1.

Intensity JND data have traditionally been expressed in many different ways, depending on the point of view of the author (6, pp. 151–152). Because there have been so many different measures, there has been a great deal of confusion as to exactly what the numbers mean. The Weber fraction was originally defined to characterize the JND where the probe and masker are identical ($\rho_{np} = 1$). When applied to maksing, J is a measure that depends on the effective correlation $\rho_e(t)$ and therefore on the temporal integration time. Until we determine how to precisely define the temporal integration time, it seems more appropriate to quote the experimental results in terms of α * rather than J, because α * does not depend on the independence assumption and therefore on $\rho_e(t)$.

Maximum Likelihood Formulation of Riesz's Experiment. When two sine waves beat, Riesz's measure $\Delta I = I_{\text{max}} - I_{\text{min}}$ is a reasonable statistic. However, we need a more general measure when dealing with arbitrary correlated critical-band signals. The method of maximum likelihood estimation is the natural way to do this (54).

One could think of Riesz's experiment either in terms of a two-hypothesis test where H_0 is for $\alpha = 0$ and H_1 is for $\alpha > 0$, or as the detection of a 3 Hz sine wave in noise where H_0 is $N(0, \sigma_{\mathscr{D}})$ and H_1 is the sine wave plus the same noise used in H_0 . This is a modulation detection (MD_I) task where the means are equal and the variance changes. Thus when calculating the probability of a 75% correct response, we cannot use the de facto rule d' = 1, because this measure is always zero (because $\Delta \mathscr{G} = 0$).

When $\rho_e = 0$, a *sufficient statistic* is the ratio of the change in mean to the variance (i.e., d'), while for the modulation detection case ($\rho_e \neq 0$), it is the variance ratio (54). Given two normal distributions N(0, 1) and $N(0, \sigma)$, the probability of correct classification by a maximum likelihood classifier is 0.742 when $\sigma = 3$. For the case of a sine wave in unit-variance noise [N(0, 1)], simulations show that $\alpha * \approx 6.0$. This then gives us a formal mechanism for relating the 2-IFC JND measurements to the modulation detection measurements.

Noise-on-Tone Maskers. The masking due to a tone and that due to a subcritical bandwidth of noise of equal intensity are very different. This difference has been called the *asymmetry of masking*. This asymmetry is clearly evident in Fig. 5 (Ref. 5, Fig. 7), which compares a five-subject average masking pattern for a 90 Hz narrow band of noise (solid curve) with the tone masking pattern. The intensity of both maskers is the same (80 dB SPL, or 65 dB SL). Fletcher (Ref. 9, p. 205) showed that the loudness of a subcritical band of noise is the same as the loudness of a pure tone having the same intensity. Even though the intensity and the loudness are the same, from Fig. 7 the masked thresholds differ by about 20 dB at the masker frequency of 400 Hz, or by about 18 dB on either side (e.g., at 380 Hz and 430 Hz).

Figures 3 and 4 in Egan and Hake's paper (5) show single-subject results at 430 Hz where $I^*_{p}/I_{m} = 1000$ (i.e., -30 dB) for the tone masker, and 100 (i.e., -10 dB) for the noise masker, leading to a 10 log 10 = 20 dB difference off the masker frequency. Their Figs. 5 and 6 provide data for a second subject at three intensities as a function of frequency. At the masker frequency of 430 Hz the difference between the noise-masked and the tone-masked threshold has a mean of 23 ± 1.6 dB.

We can use Eq. (24) to explain a significant portion of the asymmetry of masking. To do this we start with our estimate of SNR₂ from Fig. 3. For a level of 80 dB SPL at 400 Hz, we have SNR₂ \approx 40. Since $\gamma \approx \frac{1}{3}$ for tones, from Eq. (18) we estimate SNR_I to be about 13.3. From Eq. (24), with $|\rho_e| = 1$ (for tones), we have $\alpha * = 1/26.6$, or -28.5 dB. The difference between the 65 dB SL tone masker level in Fig. 5 and the dashed line at 400 Hz is about -20 dB. The difference between -28.5 dB and -20 dB (-8.5 dB) represents the error in the prediction.

For the case of the noise masker, $\rho_e = 0$. From Eq. (24), $\alpha * = 1/\sqrt{13.3} = 0.27$, or -11.2 dB. The corresponding value from Fig. 5 is the difference between the noise masker level of 65 dB SL and the 61 dB level of the solid line near 410 Hz, which is -4 dB, resulting in a -7.2 dB error. The energy detector formula gives a 28.5 - 11.2 = 17.3 dB difference between the tone and the narrowband-noise masker, compared to Egan and Hake's 20 - 4 = 16 dB difference. Thus while the absolute estimates of $\alpha *$ are too small by about a factor of two (meaning either the estimate of the loudness SNR may be too large, or the subjects were underperforming), the prediction of the asymmetry of masking is close to the measured value. There is some uncertainty in the value of the slope γ , since for noise it is $\frac{1}{4}$, while for tones it is $\frac{1}{3}$.

We conclude that the correlation between the masker and probe has a dramatic effect on the threshold signal gain α *, with threshold intensity variations of up to 32.46 dB when $SNR_I \approx 10$ [i.e., from Eq. (24) with ρ_e between -1 and 1].

The energy detector analysis clearly show the importance of the correlation between the probe and masker. When a tone is added to a narrow band of noise of the same center frequency, the two signals move slowly in and out of phase, reflecting the correlation and increasing the variance of the decision variable. We conclude that a proper analysis of masking using a maximum likelihood analysis of the detection problem, applied in the loudness domain, will result in excellent correlations with masking experiments.

The Loudness Model

When a single tone is presented to the cochlea, the energy is spread out along the cochlea, even though the energy only exists at a single frequency. The function H(f, x) defines a family of complex filter functions. Corresponding to every point x_0 there is a filter function $H(f, x_0)$, and for every pure tone at frequency f_0 , the energy is spread along the basilar membrane according to $|H(f_0, x)|^2$. Assuming the signal is above the threshold at a given point x, the excitation at each point drives nerve fibers that innervate that patch of basilar membrane. The total spike rate for that patch defines the partial loudness rate $\mathscr{D}_{tx}(t, x)$. The total loudness is given by a double integral over time and place:

$$\mathcal{L}(t) = \int_{\tau=-\infty}^{t} E_s(t-\tau) \int_{x=0}^{1} \mathcal{L}_{tx}(\tau, x) \, dx \, d\tau$$

where $E_s(t)$ is Munson's (30) sensation integral, and the integral over x is normalized to unit length. The partial loudness function $\mathscr{D}_{tx}(t, x)$ is a nonlinear transformation of the energy along the basilar membrane, defined by the bank of filters $H(f, x) = \mathscr{F} \cdot h(t, x)$ [\mathscr{F} represents the Fourier transform, and h(t, x) is a family of impulse responses].

Some Basic Questions

Here are some basic questions that are begging for further investigation:

- Why is the critical ratio independent of level (9,25,57)?
- What is the relation between masking by narrow bands of frozen noise and the pure-tone JND?
- Are Weber's law and the linear relation seen in narrowband and wideband maskers related?
- Why does the masked threshold track the masker intensity in a linear manner over such a large range of intensities, given that the BM at CF is nonlinear? That is, why is $\sigma_I \propto I$?
- Why do we find Weber's law to hold for wideband noise (29)?
- What is the reason for the near miss to Weber's law for tonal signals (6,29,44)?
- What is the source of the upward spread of masking (1,8,58)?
- Why is there a > 26 dB difference in the masking between a tone and a narrow band of noise (5)?
- What determines the ratio of the critical ratio to the ERB, and what is its true value?
- What are the relative contributions of internal and external noise in the masking of tones and narrow bands of noise, as a function of the bandwidth and duration of the signal?
- Why is there such a close relation between the frequency JND and the ERB (8,9)?
- Does loudness add (18)?
- What is the loudness SNR for two equally loud tones that do not mask each other?

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